# StateNon-9-90 V0-90 V18 A0-90 V0-90 V19 A0-90 V19 A19 A0-90 V

The MCOTS NiQor<sup>®</sup> Quarter-brick dc-dc converter is a non-isolated buck-boost regulator, which employs synchronous rectification to achieve extremely high conversion efficiency. The High Input Voltage MCOTS NiQor Brick family of converters can be used in traditional DPA (distributed power architecture) systems or provide a regulated output voltage from a battery source or other variable voltage source. The MCOTS NiQor Brick family can be configured to Buck the input voltage down to a lower voltage or Boost the input voltage up to a higher voltage using a single external resistor.

#### **Operational Features**

- High efficiency, 96% at full rated load current
- Delivers up to 18 A of output current
- Input Voltage Range: 9-90 Vdc
- Output Voltage Range: 0-90 V (negative output is possible)
- Extensive on-board input and output filtering
- No minimum load requirement means no preload resistors required
- Adjustable current limit with current monitor

## **Protection Features**

- Input under-voltage lockout protects the converter at low input voltage conditions
- Over-current shutdown protects converter from excessive load current or short circuits
- Input/output over-voltage protection protects load and regulator from damaging voltages
- Thermal shutdown protects converter from abnormal environmental conditions

# Screening/Qualification

- AS9100 and ISO 9001 certified facility
- Qualified to MIL-STD-810
- Available with S-Grade or M-Grade screening
- Pre-cap inspection per IPC-A-610, Class III
- Temperature cycling per MIL-STD-883, Method 1010, Condition B, 10 cycles
- Burn-In at 100 °C baseplate temperature
- Final visual inspection per MIL-STD-883, Method 2009
- Full component traceability





## Designed and manufactured in the USA.

# **Control Features**

- On/Off control
- Output voltage trim permits custom voltages
- Settable current limit
- Output voltage trim range of 0V 90V

# **Mechanical Features**

- Industry standard pin-out configuration
- Standard size: 1.54" x 2.39" x 0.515"

39.01 x 60.6 x 13.08 mm

• Total weight: Encased - 3.5 oz. (100 g)

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**Current: 18A** 

pecifications subject to change without notice.		1	1	1		
Parameter	Vout	Min.	Тур.	Max.	Units	Notes & Conditions
ABSOLUTE MAXIMUM RATINGS						
input Voltage						
Non-Operating	All	-1		100	V	Continuous
Operating	All			90	V	Continuous
solation Voltage						
Input to Baseplate				1150	Vdc	
Output to Baseplate				1150	Vdc	
Storage Temperature	All	-65		125	°C	
/oltage at ON/OFF input pin	All	0		5.5	V	
/oltage at Vset and Iset Pins		-0.2		3.5	V	Note 1
RECOMMENDED OPERATING CONDITIONS						
input Voltage Range	All	9		90	V	Turn on at 10V
input Fuse Rating	All			30	A	Fast blow external fuse recommended
nput Current				18	A	Input current max equals output rated current
External Input Capacitance	All	680			μF	ESR > 50 m $\Omega$ See Note 2
Dutput Voltage	All	0		90	V	
Dutput Current	All	0		18	Α	Input voltage dependent
INPUT CHARACTERISTICS						
nput Under-Voltage Lockout						
Turn-On Voltage Threshold	All	9.2	9.5	10	V	
Turn-Off Voltage Threshold	All	5.1	5.5	5.9	V	
Lockout Hysteresis	All		4.0		V	
nput Current Limit	All			20.3	A	
lo-Load Input Current	12		115		mA	
w	48		210		mA	
W	72		115		mA	
Disabled Input Current	All		1.0		mA	
nput Filter Components Value (C\L\C)	All		13 \ 0.33 \ 40		µF\µH\µF	
DUTPUT CHARACTERISTICS						
Dutput Voltage Range	All	0		90	V	Set by Vset resistor
Dperating Output Current Range	All	0		18	Α	
Dutput Voltage Regulation						
Load Regulation	All	-2%	*Vout*Iout/In	nax		
Total Output Voltage Range	All	±100mV ±	⊧2%*Vout + I	oad Reg		Between Output pins, over sample, line, load, temp. & life
Output Voltage Ripple and Noise (pk-pk/28Vin)	12		50		mV	48 Vin Full Load; 100uF; 20 MHz b.w.
n n	48		200		mV	<i>n</i>
"	90		200		mV	w
Output DC Over Current limit	All		20.3		A	Effective on input, output condition
External Output Capacitance	All	100			μF	$\text{ESR} > 1 \text{ m}\Omega$
Dutput Filter Components	All		40 \ 0.33 \ 10		μF\μH\μF	
Reverse Current	All		1		μΑ	Disabled
DYNAMIC CHARACTERISTICS			1			l 
Dutput Voltage during Current Transient						
For a Step Change in Output Current	12		1100		mV	(0.1 A/µs); 50%-75%-50% Iout max
Settling Time	12		300		us	To within 1.5% Vout nom.
For a Step Change in Output Current	48		1400		mV	(0.1 A/µs); 50%-75%-50% Iout max
Settling Time	48		1000		us	To within 1.5% Vout nom.
-	10		1000		45	
urn on Transient						
	ΔII		2		mc	Resistive load
Furn on Transient Startup delay Rise Rate	All All		2 2.3		ms V/ms	Resistive load

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Parameter	Vout	Min.	Typ.	Max.	Units	Notes & Conditions
EFFICIENCY						
100% Load; 48 Vin	12		90		%	
100% Load; 72 Vin	48		96		%	
100% Load; 12 Vin	72		88		%	
50% Load; 48 Vin	12		91		%	
50% Load; 72 Vin	48		96		%	
50% Load; 12 Vin	72		89		%	
FEATURE CHARACTERISTICS						
Switching Frequency	All	240	250	260	KHz	
On/Off, Negative (N) Logic						See REMOTE ON/OFF: in CONTROL FEATURES
Off-State Threshold Voltage	All		1.4	1.5	V	
On-State Threshold Voltage	All	1.0	1.1		V	
Threshold Hysteresis	All		0.3		V	
Pin Pull-Up Voltage	All		5		V	
Pin Pull-Up Resistance	All		25		kΩ	
Output Voltage Setpoint						See OUTPUT VOLTAGE SETPOINT: in CONTROL FEATURE
Pin Pull-Up Voltage	All		2.5		V	
Pin Pull-Up Resistance	All		10.9		kΩ	
Output Voltage Setpoint Range	All	0		90	V	
Output Over-Voltage Shutdown	All		104		V	Fixed - Does vary with output setpoint
Output Current Setpoint						See OUTPUT CURRENT SETPOINT: in CONTROL FEATUR
Pin Pull-Up Voltage	All		2.5		V	
Pin Pull-Up Resistance	All		10		kΩ	
Output Current Setpoint Range	All	0		20.2	A	
Input/Output Current Limit	All	18.4	20.2	22.0	A	
Ishare/Imon						See OUTPUT CURRENT SHARE: in CONTROL FEATURES
Pin Voltage at No Load	All		0.2		V	
Pin Voltage at Full Load (Imax)	All		2.2		V	
Pin Output Resistance	All		2.5		kΩ	
Over-Temperature Shutdown	All		115		°C	Average PCB Temperature
Over-Temperature Shutdown Restart Hysteresis	All		15		°C	
RELIABILITY CHARACTERISTICS			'	·	·	
Calculated MTBF (TR-NWT-000332; Telcordia)	All		3.5		10 <sup>6</sup> Hrs.	70 °C Tbaseplate
Calculated MTBF (MIL-HDBK-217F; MIL-217)	All		3.2		10 <sup>6</sup> Hrs.	70 °C Tbaseplate
Field Demonstrated MTBF	All				10 <sup>6</sup> Hrs.	See our website for details
TEMP LIMITS FOR POWER DERATING						
Semiconductor Junction Temperature	All			125	°C	Package rated to 150 °C
Board Temperature	All			125	°C	UL rated max operating temp 130 °C

Note 1: All control signals are referenced to Vout- pin.

Note 2: Input capacitance and ESR of input capacitance dictated by input stability requirements. For Vin >> Nom Vin and/or Vout >> Nom Vout, consult SynQor Field Application

Please see SynQor Application Note: http://www.synqor.com/documents/appnotes/appnt\_System\_Instability.pdf

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# MCOTS-N-28VE-90-QT

**Output: 0 - 90V Current: 18A** 

# **Technical Charts**



Figure 1: Efficiency at different output voltages vs. load current for different input voltages at 25°C.



Figure 3: Maximum output power derating curve with a controlled baseplate temperature of 80°C and 100°C vs. Input voltage.



Figure 5: Input Terminal Ripple Current at 28V input and rated load current (1A/ div). Load capacitance: 100uF electrolytic cap. Bandwidth: 20MHz, (2uS/div). See Figure 4



Figure 2: Power dissipation at different output voltages vs. load current for different input voltages at 25°C.



Figure 4: Test set-up diagram showing measurement points for Input Terminal Ripple Current (Fig 5) and Output Voltage Ripple (Fig 6).



Figure 6: Output Voltage Ripple at 28V input and rated load current (200mV/ div). Load capacitance: 100uF electrolytic cap. Bandwidth: 20MHz, (2uS/div). See Figure 4.

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**Technical Charts** 



Figure 7: Turn-on transient at 28V input; full load (10ms/div). Top Traces: Vout (20V/div). Bottom Trace: ON/OFF input (5V/div)



Figure 9: Vset pin voltage vs. Output Voltage Setpoint.



Figure 11: Iset pin voltage vs. Output Current Limit Setpoint.







Figure 10: Output Voltage vs. Vset pin voltage dynamics; 48Vin, 10A out (10mS/ div). Bottom trace: Vset pin voltage (500mV/div). Top trace: Output Voltage (10V/div).



Figure 12: Output Current vs Iset pin voltage dynamics; 48Vin 12Vout (2mS/div). Top Trace: Output Current (5A/div). Bottom Trace: Iset pin voltage (500mV/div).

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**Technical Charts** 



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Figure 13: Ishare/Imon pin Voltage vs. Output Load Current.



Figure 14: Output voltage response for 48V input; 12V, 28V, 48V, 90V output to step change in load current (50%-75%-50% of lout max; di/dt=0.1A/uS). Load cap: 100uF electrolytic cap, Vout (1V/div); (1mS/div)

## STANDARDS COMPLIANCE

Parameter	
STANDARDS COMP	Π

Notes & Conditions

#### STANDARDS COMPLIANCE UL 62368-1

EN 62368-1

CAN/CSA-C22.2 No. 62368-1

Note: An external input fuse must always be used to meet these safety requirements.

Contact SynQor for official safety certificates on new releases or download from the SynQor website.

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# **Application Section**

#### **BASIC OPERATION AND FEATURES**

This converter consists of integrated buck and boost converters, both controlled simultaneously by a digital controller. It automatically changes operating mode (buck mode or boost) when the line voltage or output set point changes. Very high efficiency is maintained over wide input and output ranges by shifting operational modes and use of synchronous rectifiers.

The converter runs at a fixed frequency with a predictable EMI performance.

This quarter-brick converter uses the industry standard footprint and pin-out configuration. A typical V-I characteristic, with Vsetpoint=60V and Isetpoint=15A, when operating from Vin=45V, is shown in Figure A.





#### **CONTROL FEATURES**

**REMOTE ON/OFF:** Only Negative On/Off logic is available in the converter series: logic high at the input turns the converter Off while a logic low turns in On. Timing of this is shown in Figures 7, 8. A high level can be driven to any voltage between 1.5V and 3.3V, or simply left floating as the unit contains an internal  $25K\Omega$  pull-up to 5.0V. The pin can be pulled low (to Vout-) by an optocoupler, an open-source/ drain transistor or wired permanently to Vout-. To provide noise immunity the input has 0.3V of hysteresis.

Multiple units that have a common Vout- connection can be controlled by the same On/Off signal, but it is recommended that a small schottky diode be added to each input as shown in Figure B.



Figure B: Negative output setup

**OUTPUT VOLTAGE SETPOINT:** The output voltage can be programmed to any voltage between 0 V dc and Vmax by connecting one resistor between the Vset pin (6) and Vout- (4); See Figure C. For a desired output voltage, the value of the resistor should be:

$$Rvset(Vset) = \left[ \left( \frac{11830 \times Vmax}{Vset + 0.058 \times Vmax} \right) -10912 \right] (\Omega)$$

# **Application Section**



#### Figure C

Alternatively, the Vset pin can be driven from an external voltage source: Undriven, this pin floats at 2.5V which sets the output to 0V. See Figure 10 for the large scale dynamics of this input.

See Figure 9

where:

Vset = desired output voltage setpoint

Vmax = maximum rated output voltage (90V)

Vset

**OUTPUT CURRENT SETPOINT:** The maximum output current (effectively the current limit) can be reduced to any value between 0 and Imax by connecting one resistor between the Iset pin (5) and Vout- (4); see Figure C. The value of the resistor should be:

RIset(Iset) =	(	0.0469 Imax + Iset	)	*10200 -10	٦	<b>(Ω</b> )
L		1.153 Imax - Iset	)			()

Alternatively, the Iset pin can be driven from an external voltage source:

$$Viset(Iset) = (0.0953+2.085 * Iset/Imax) V$$

where:

Iset = desired output current setpoint Imax = maximum rated output current (18A)

Undriven, the Iset pin floats to 2.45V which sets the current limit at its nominal value of 112.5%\*Imax. See Figure 12 for the large signal dynamics of this control.

#### **PROTECTION FEATURES**

**Input Under-Voltage Lockout:** The converter is designed to turn off when the input voltage is too low, helping avoid an input system instability problem, described in more detail in the application note titled "Input System Instability". The lockout circuitry is a comparator with DC hysteresis. When the input voltage is rising, it must exceed the typical Turn-On Voltage Threshold value (listed on the specification page) before the converter will turn on. Once the converter is on, the input voltage must fall below the typical Turn-Off Voltage Threshold value before the converter will turn off.

**Output Current Shutdown:** To provide protection in an output short condition, the unit is equipped with internal short circuit protection. When the short-circuit protection is triggered, the converter shuts down and then waits an inhibit time ( $\sim$ 100ms), after which it tries to turn on again. If the short condition remains, the current limit circuit will limit the output current. The unit will return to normal operation once the fault condition is removed.

**Internal Over-Voltage Protection:** To fully protect from excessive output voltage, the unit contains an output over-voltage shutdown that is fixed at  $\sim 110\%$  of Vmax. If this limit is reached, the converter shuts down and then waits an inhibit time ( $\sim 100$ ms), after which it restsarts.

**Over-Temperature Shutdown:** Two sensors in the module monitor the temperature of both the buck and boost sections. When the temperature at either sensor exceeds the Over Temperature Shutdown value, the converter is disabled. It will restart normally when it then cools by the amount of the Over-Temperature Shutdown Hysteresis.

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# **Application Section**

#### **APPLICATION CONSIDERATIONS**

**Input filtering:** These modules should be connected to a low-impedance source. A highly inductive source can affect the stability of the module. An input capacitor must be placed adjacent to the input pins of the module to minimize input ripple voltage and ensure stability. See SynQor's application note titled Input System Stability for guidance on selecting appropriate input filter elements.

#### http://www.synqor.com/documents/appnotes/appnt\_System\_Instability.pdf

**Output capacitance:** An output capacitor of at least 100uF with an ESR of ~50mOhms is recommended to damp the resonance of the output filter. More capacitance can be added to improve dynamic response to load variation; there is not upper limit on the value of the output capacitance. Very large output capacitors may slow the converter's start up time due to the current required to charge them being limited by current limit or Isetpoint.

**Current limit:** Available current is limited by maximum output current in buck mode (Vin>Vout), and by maximum input current in boost mode (Vin<Vout).

for Vin >Vout

Ilimit = 1.12 x Imax (Amps)

for Vin < Vout

where Imax = Maximum rated output current

Ilimit =  $1.12 \times \text{Imax} \left( \frac{\text{Vin}}{\text{Vout}} \right)$  (Amps)

**Generating Negative output voltage (inversion):** The circuit shown in Figure D can be used to generate a negative output voltage from a positive input.



Figure D: Negative output setup

Note that all control signals are referenced to Vout-.

# **Application Section**

**Thermal Consideration:** The maximum operating baseplate temperature, TB, is 100 °C. Refer to the Thermal Derating Curves in the Technical Figures section to see the available output current at baseplate temperatures below 100 °C.



A power derating curve can be calculated for any heatsink that is attached to the base-plate of the converter. It is only necessary to determine the thermal resistance, RTHBA, of the chosen heatsink between the baseplate and the ambient air for a given airflow rate. This information is usually available from the heatsink vendor. The following formula can be used to determine heatsink vendor. the maximum power the converter can dissipate for a given thermal condition if its base-plate is to be no higher than 100 °C.

This value of maximum power dissipation can then be used in conjunction with the data shown in the Power Dissipation Curves in the Technical Figures section to determine the maximum load current (and power) that the converter can deliver in the given thermal condition. For convenience, Thermal Derating Curves are provided in the Technical Figures section.

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# Standard Mechanical Diagram





#### NOTES

- 1) Applied torque per screw should not exceed 6in lb (.7Nm)
- 2) Screw depth is not to exceed 0.100" (2.54mm) below the surface of the baseplate
- 3) Baseplate flatness tolerance is 0.004" (.10mm) TIR for surface
- 4) Pins 1-3 and 5-7 are 0.040" (1.02mm) dia. with 0.080" (2.03mm) dia. standoff shoulders
- 5) Pins 4 and 8 are 0.062" (1.57mm) dia. with .100" (2.54mm) dia. standoff shoulders
- 6) Undimensioned components are shown for visual reference only
- 7) All pins: Material: Copper Alloy, Finish: Matte Tin over Nickel Plate
- 8) All dimensions in inches[mm]. Tolerances: X.XXIN +/-0.02 (X.Xmm +/-0.5mm) X.XXXIN +/-0.010 (X.XXmm +/-0.25mm)
- 9) Weight: 3.5 oz. (100 g)
- 10) Workmanship: Meets or exceeds IPC-A-610 Class II

#### **PIN DESIGNATIONS**

Pin	Label	Function
1	+Vin	Positive Supply Input
2	On/Off	Input to enable/disable the converter, TTL
3	-Vin	Negative Supply Input, internally connected to Pin 4
4	-Vout	Negative Power Output, internally connected to Pin 3
5	lset	Input to set the maximum output current.
6	Vset	Input to set the maximum output voltage.
7	Imon	Imon
8	+Vout	Positive Power Output

# **Flanged Mechanical Diagram**





#### NOTES

- 1) APPLIED TORQUE PER M3 OR 4-40 SCREW SHOULD NOT EXCEED 6in-lb (0.7Nm).
- 2) BASEPLATE FLATNESS TOLERANCE IS 0.010" (.25mm) TIR FOR SURFACE.
- PINS 1-3 and 5-7 ARE 0.040" (1.02mm) DIA. WITH 0.080" (2.03mm) DIA. STANDOFFS.
- 4) PINS 4 AND 8 ARE 0.062" (1.57mm) DIA. WITH 0.100" (2.54mm) DIA STANDOFFS
- 5) ALL PINS: MATERIAL: COPPER ALLOY FINISH: MATTE TIN OVER NICKEL PLATE
- 6) WEIGHT: 3.78 oz. (107 g) TYPICAL
- 7) ALL DIMENSIONS IN INCHES(mm)
- 8) TOLERANCES: X.XXIN +/-0.02 (X.Xmm +/-0.5mm) X.XXXIN +/-0.010 (X.XXmm +/-0.25mm)

#### **PIN DESIGNATIONS**

Pin	Label	Function
1	+Vin	Positive Supply Input
2	On/Off	Input to enable/disable the converter, TTL
3	-Vin	Negative Supply Input, internally connected to Pin 4
4	-Vout	Negative Power Output, internally connected to Pin 3
5	Iset	Input to set the maximum output current.
6	Vset	Input to set the maximum output voltage.
7	Imon	Imon
8	+Vout	Positive Power Output

# **Qualifications & Screening**

#### **Mil-COTS Qualification**

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Test Name	Details	# Tested (# Failed)	Consistent with MIL-STD-883F Method
Life Testing	Visual, mechanical and electrical testing before, during and after 1000 hour burn-in @ full load	15 (0)	Method 1005.8
Shock-Vibration	Visual, mechanical and electrical testing before, during and after shock and vibration tests	5 (0)	MIL-STD-202, Methods 201A & 213B
Humidity	+85 °C, 95% RH, 1000 hours, 2 minutes on / 6 hours off	8 (0)	Method 1004.7
Temperature Cycling	500 cycles of -55 °C to +100 °C (30 minute dwell at each temperature)	10 (0)	Method 1010.8, Condition A
Solderability	lerability 15 pins		Method 2003
DMT	•65 °C to +110 °C across full line and load specifications in 5 °C steps		
Altitude	70,000 feet (21 km), see Note	2 (0)	

Note: A conductive cooling design is generally needed for high altitude applications because of naturally poor convective cooling at rare atmospheres.

#### Mil-COTS Converter and Filter Screening

Screening	Process Description	S-Grade	M-Grade
Baseplate Operating Temperature		-55 °C to +100 °C	-55 °C to +100 °C
Storage Temperature		-65 °C to +135 °C	-65 °C to +135 °C
Pre-Cap Inspection	IPC-A-610, Class III	•	•
Temperature Cycling	MIL-STD-883F, Method 1010, Condition B, 10 Cycles		•
Burn-In	100 °C Baseplate	12 Hours	96 Hours
Final Electrical Test	100%	25 °C	-55 °C, +25 °C, +100 °C
Final Visual Inspection	MIL-STD-883F, Method 2009	•	•

Mil-COTS MIL-STD-810G Qualification Testing							
MIL-STD-810G Test	Method	Description					
Fungus	508.6	Table 508.6-I					
Altitude	500.5 - Procedure I	Storage: 70,000 ft / 2 hr duration					
Altitude	500.5 - Procedure II	Operating: 70,000 ft / 2 hr duration; Ambient Temperature					
Rapid Decompression	500.5 - Procedure III	Storage: 8,000 ft to 40,000 ft					
Acceleration	513.6 - Procedure II	Operating: 15 g					
Salt Fog	509.5	Storage					
High Tomporature	501.5 - Procedure I	Storage: 135 °C / 3 hrs					
High Temperature	501.5 - Procedure II	Operating: 100 °C / 3 hrs					
	502.5 - Procedure I	Storage: -65 °C / 4 hrs					
Low Temperature	502.5 - Procedure II	Operating: -55 °C / 3 hrs					
Temperature Shock	503.5 - Procedure I - C	Storage: -65 °C to 135 °C; 12 cycles					
Rain	506.5 - Procedure I	Wind Blown Rain					
Immersion	512.5 - Procedure I	Non-Operating					
Humidity	507.5 - Procedure II	Aggravated cycle @ 95% RH (Figure 507.5-7 aggravated temp - humidity cycle, 15 cycles)					
Random Vibration	514.6 - Procedure I	10 - 2000 Hz, PSD level of 1.5 g <sup>2</sup> /Hz (54.6 $g_{rms}$ ), duration = 1 hr/axis					
Shock	516.6 - Procedure I	20 g peak, 11 ms, Functional Shock (Operating no load) (saw tooth)					
SHOCK	516.6 - Procedure VI	Bench Handling Shock					
Sinusoidal vibration	514.6 - Category 14	Rotary wing aircraft - helicopter, 4 hrs/axis, 20 g (sine sweep from 10 - 500 Hz)					
Sand and Dust	510.5 - Procedure I	Blowing Dust					
Sanu anu Dust	510.5 - Procedure II	Blowing Sand					

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## **Ordering Information/ Part Numbering**

Example: MCOTS-N-28VE-90-QT

Not all combinations make valid part numbers, please contact SynQor for availability. See product summary page for details.

Family	amily Product Input Voltage		nily Product Input Voltage				Product Input Voltage Output Voltage Package		Thermal Design	Screening Level	
мсотя	N: Non-Isolated	<b>28V:</b> 9-60V <b>28VE:</b> 9-90V	<b>60:</b> 0-60V <b>90:</b> 0-90V	<b>QT:</b> Quarter Brick Tera	N:Normal Threaded F:Flanged	<b>S:</b> S-Grade <b>M:</b> M-Grade					

#### **Application Notes**

A variety of application notes and technical white papers can be downloaded in pdf format from our website.

#### Contact SynQor for further information and to order:

 Phone:
 978-849-0600
 Fax:
 978-849-0602

 E-mail:
 power@synqor.com
 Web:
 www.synqor.com

 Address:
 155
 Swanson Road, Boxborough, MA 01719
 USA

#### WARRANTY

SynQor offers a two (2) year limited warranty. Complete warranty information is listed on our website or is available upon request from SynQor.

#### PATENTS

SynQor holds numerous U.S. patents, one or more of which apply to most of its power conversion products. Any that apply to the product(s) listed in this document are identified by markings on the product(s) or on internal components of the product(s) in accordance with U.S. patent laws. SynQor's patents include the following:

7,765,687 7,787,261 8,149,597 8,644,027